Three-Phase Five-Level Flying Capacitor Multilevel inverter For Harvesting Solar Power

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I. Introduction

There Has Been A Steady Increase In The Demand For Electricity Because Of Increase In Energy Demand Due To Increase In Population, Economy Growth And The Rapid Depletion Of Fossil Fuel Sources Based On Energy Reserve And Rapid Growth Of Energy Demand. The Research For An Alternative Source Of Power Generation Has Resulted In Renewable Energy As Source Of Energy Which Possibly Has No Harm On The Environment (Chauhan Et Al., 2013) And Is Abundant Especially Tropic Region Of The World. In 2000, Ieee Defines An Inverter As A Device That Converts Electrical Power From A Dc Voltage (De-Source) Into A Symmetric Ac Voltage Output Of Desired Magnitude And Frequency Using Electronic Circuits. Numerous Industrial Applications Have Begun To Require Higher Power Apparatus In Recent Years. Some Medium Voltage Motor Drives And Utility Applications Require Medium Voltage And Megawatt Power Level. For Medium Voltage Grid, It Is Troublesome To Connect Only One Power Semiconductor Switch Directly. As A Result A Multilevel Power Inverter Structure Has Been Introduced As An Alternative In High Power And Medium Voltage Situations. A Multilevel Inverter Not Only Achieves Higher Power Ratings, But Also Enables The Use Of Renewable Energy Sources, Such As Photovoltaic, Wind And Fuel Cells Can Easily Interfaced To A Multilevel Inverters System For Higher Applications In Rodriguez Et Al. (2002). Many Researches Have Been Done On Solar Energy For Grid Connection, They Are:


At Near-Unity Power Factor.


An Investigation Study Of Total Harmonic Distortion In A Flying Capacitor Multilevel Inverter With/Without Closed–Loop Feedback Schemes Was Proposed By Shanmuga Et Al. (2013). They Used A Seven Level To Overcome The Main Drawbacks Of These Inverters Which Are High Harmonic Content And Used Only For Limited Power Applications. In Order To Overcome This, A Novel Approach Called Closed-Loop Fcmli Was Proposed, Which Significantly Increases The Level Number Of The Output Waveform And Thereby Dramatically Reduces The Low-Order Harmonics And Thd. The Proposed System Consists Of A Dc–Dc Power Converter And A Dc–Ac Multilevel Inverter. In Order To Achieve Low Cost, Easy Control, High Efficiency And High Reliability, A Capacitor Clamped Dc–Dc Boost Converter Using Minimal Devices Was Introduced To Interface The Low Voltage Photovoltaic (Pv) Module.

II. Materials And Methods

Figure 1 Shows The Block Diagram Of The Three-Phase Controlled Load Using Five-Level Flying Capacitor Multilevel Inverter Which Consists Of subsections That Made Up The Whole System. The Array Of Solar Pv Are Arranged In Series And Parallel To Generate The Required Current And Voltage. The Solar Pw Array Is Connected To A Current/Voltage Controller That Keeps The Current And Voltage Within A Rated Current Range, So That Outside This Rated Current And Voltage There Will Be No Output From The Current/Voltage Controller To Switch The Mosfet Of Boost Converter. The Boost Converter Duty Circle Is Controlled By The Rated Current And Voltage Of The Solar Pv And Boost Converter Is Turned To Give A Constant Output Through The Pi. The Step Up Voltage From The Boost Converter Is Fed Into The Three-Phase Five-Level Fc Mli Using Pod Switching Pattern,An Ac Voltage Output Is Generated From The Three-Phase Five-Level Fc Mli Which Is Connected To Three Phase Load (Grid) Via Filter And Transformer.

2.1 Pv Sizing Determine Power Consumption Demands

The First Step In Designing The Array Of The Solar Pw System Is To Find Out The Total Power And Energy Consumption Of All Loads That Needed To Be Connected To The Solar Pw System. In This Paper The Solar Pw Panel Is Expected To Power A Load Of 240v, 15a. Since We Are Using 100w, 12vdc Solar Pw Panel To Generate The Voltage Range Of 200v To 240v And Current Of 12a To 15a, We Calculate The Current Using
Equation (1).
\[ P = V_i \]  
(1)

Current = 100/12 = 8.8a
240v = Connecting Ten (20) Pvs In Series Gives
16.6a = Connecting This Ten (2) Pvs In Parallel Gives

2.2 Boost Converters

The Boost Converter Circuit Diagram Is Shown In Figure 2(A). This Is Another Switching Converter That Operates By Periodically Opening And Closing An Electronic Switch. It Is Called A Boost Converter Because The Output Voltage Is Larger Than The Input. The Boost Circuit Consists Of Energy Storing Element Inductor, A Capacitor, A Diode, A Load And A Switching Device Like Mosfet. The Equation For Voltage Step Up To Occur Is Shown In Equation 2. (Muhammad, 2004)

\[ \text{Duty ratio}(D) = 1 - \frac{V_i}{V_o} \]  
(2)

\[ (L)_{min} = \frac{D(1-D)^2R}{2f} \]  
(3)

\[ C \geq \frac{R(\Delta V_o/V_o)}{f} \]  
(4)

The Boost Converter Is Designed To Give An Output Voltage Of 415vdc From Source Voltage Of 200-240vdc; With Continuous Inductor Current, An Output Ripple Voltage Of Less Than 1%, Load Resistance Of 50ω And Switching Frequency Of 5khz, The Values Of The Inductor And Capacitor Can Be Obtain By Using Equations 2, 3 And 4.

The Block Diagram Of The Three-Phase Five-Level Fc Mli Is Shown In Figure 3. Using Pod Technique (Bin Wu, 2006) At The Control Block Which Is Essentially The Pwm Circuits That Consist Of An Aggregation Of A Reference Sine Wave Generator (Phase Shift Oscillator), A Carrier Signal Triangular Wave Generator (Integrator) And The Comparator (Modulator). This Arrangement Generates Both The Pwm Reference And Carrier Train Of Pulses And Also Compares (Modulates) Same In Order To Generate The Pwm Output That Is Then Transmitted To The Power Circuit To Trigger And Control The Switching Action Of The H-Bridge Modules. The Three-Phase Five-Level Fcmli Consists Of Three Sinewaves Of Each Displayed By A Phase Shifts Of 120 Degrees Each And Four Triangular Waves Of 0 To 1 And 1 To 2, 0 To -1 And -1 To -2 Amplitude. These Four Triangular Waves Are Fed Into A Op-Amp That Compares Each Of The Sine Waves (0, 120, And 240) And The Output Of Each Of This Comparison Are Noted To Give Eight Pwm Signals For Each Phase (0, 120, And 240) Which Are Fed Into Each Phase Of The Three Phase Five-Level Five Fc Mli. The Outputs Of The Three-Phase Five-Level Fcmli Are Filtered And Then Connected To Three Phase Isolation Transformer Before The Output From The Transformer Is Connected To A Grid.
Figure 3: Block Diagram Of The Five Level Three Phase Fc Mli

The Circuit In Figure 4 Shows A Single Phase Five-Level Fcmli Which Uses Eight Power Semiconductor Switches, Three Flying Capacitors And Two Dc Link Capacitors. This Fcmli Consists Of Four Switching Pairs (S1 S8), (S2 S7), (S3 S6) And (S4 S5). If One Switch Of The Pair Is Switched On, The Other Complementary Switch Of Same Pair Must Be Off. The Switches Are Clamped By Dc-Link Together With Flying Capacitors. The Four Switches (S1-S4) Must Be Connected In Series Between Dc Input And Load And Likewise For (S5-S8). The Three Flying Capacitors C3, C4 And C5 Are Charged To Different Voltage Levels. By Changing The Transistor Switching States, The Capacitors And The Dc Source Are Connected In Different Ways To Produce Various Load Voltages. Typical Switch Combinations To Obtain The Required Output Voltage Levels For Five-Level Fcmli Are Shown In Table 1. In The Circuit Configuration Of The Three-Phase Five-Level Fc Mli, The Inverter Is Powered By A Single Dc Source. The Dc Source Is Obtained From The Output Of The Boost Converter.

Figure 4: Basic Circuit Of A Single-Phase Five-Level Fc Mli.

Table 1: Switching States/Patterns Of A Five-Level Fc Mli (Positive Load Current)

<table>
<thead>
<tr>
<th>Output Voltage Level</th>
<th>S1</th>
<th>S2</th>
<th>S3</th>
<th>S4</th>
<th>S5</th>
<th>S6</th>
<th>S7</th>
<th>S8</th>
</tr>
</thead>
<tbody>
<tr>
<td>Vd/2</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>Vd/4</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>Vd/8</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
</tbody>
</table>
Three-Phase Five-Level Flying Capacitor Multilevel Inverter For Harvesting Solar Power

<table>
<thead>
<tr>
<th>Vdc/4</th>
<th>1</th>
<th>1</th>
<th>0</th>
<th>1</th>
<th>0</th>
<th>1</th>
<th>0</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Vdc/4</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>Vdc/4</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>Vdc/4</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
</tbody>
</table>

Figure 5 Shows The Generation Of Eight PWM For Phase 0° Using Sinewave Of 0° Phase Shift To Combine With Four (4) Triangular Waveform Generators Of Amplitude Of 0 To 1, 1 To 2, 0 To -1 And -1 To -2. The Result Of The Comparing Gives Four PWM; Which When Noted Gives Eight PWM. The Same Procedures Go For Figure 6 And 7 Where The Sinewave Generators Are 120° And 240° Respectively. The Output Of These Twenty-Four PWM Is Fed Into The Twenty-Four Switches (MOSFET) Of The Three-Phase Five-Level FCMLI To Generate A Modified Sinewave AC Voltage Output Of Five Steps. The Three-Phase Five-Level FCMLI Is Shown In Figure 8. The Logic Conditions For The Switching MOSFET Is Shown In Equation 5 To 8. Vt Is The Sine Wave Voltage And Vm1 To Vm4 Are Voltages For The Four Triangular Wave Of, 2 To 1, 1 To 0, 0 To -1 And -1 To -2 Respectively. The Logic Conditions For The Eight MOSFET To Switch Are:

If Vm1 > Vt Then A State Is On; If Vm1 < Vt Then Not A State Is On
(5)
If Vm2 > Vt Then B State Is On; If Vm2 < Vt Then Not B State Is On
(6)
If Vm3 > Vt Then C State Is On; If Vm3 < Vt Then Not C State Is On
(7)
If Vm4 > Vt Then D State Is On; If Vm4 < Vt Then Not D State Is On
(8)

Figure 5: SIMULINK Control Circuit for Phase One at 0°

Figure 6: SIMULINK Control Circuit for Phase One at 120°

Figure 7: SIMULINK Control Circuit for Phase One at 240°

Figure 8: Circuit configuration of the three-phase five-level FC MLI

Figure 9 Shows The Complete Circuit Diagram Of The Three-Phase Five-Level Flying Capacitor Multilevel Inverter For Harvesting Solar Power. The Circuit Is Segmented Into A To G. Section A Consist Of The Solar Panel Array That Is Arrange In Series And Parallel To Generates The 240v, 15a. The Solar Array Was Connected To The Current/Voltage Controller Circuit, Which Is The Section B. The Current/Voltage Controller...
Circuit Make Sure The Current And Voltage Is Within The Rated Current And Voltage Value Of 12-17a And 200 To 240v. The Output From Section B Is Connected To The Mosfet Of Section C. Section C Is The Boost Converter, That Step Up The Voltage Range Of 200-240vdc To 415vdc. The Output From The Boost Converter Serve As The Voltage Source To Section D. Section D Is The Three-Phase Five-Level Mli. Is Made Up Of Twenty-Four Switches; The Inverter Converts The 415vdc To 415vac. The Three-Phase Five-Level Mli Is Controlled By Section E. Section E Is The Pwm Switching Generators Of The Mli Using Pod Switching Technique. Section F Comprises Of The Filter Connected To Transformer And The Transformer Connected To The Grid. Section G Consist Of Current And Voltage Measurement Circuits Where Display Readings Are Taken.

III. Results And Discussions

Figure 10 Shows The Solar Pv Connected To Current/Voltage Controller Circuit Which Output Is Connected To The Boost Converter Mosfet. The Current And The Voltage Of The Solar Pv Is Being Measured By The Current/Voltage Controller Circuit. The Allowed Current For Dc Voltage Step Up To Occur Is 12a To 17a And The Equivalent Scaled Voltage Value Is 3.8v To 5v Using Voltage Divider Rule Where R1 = 1k And R2 = 43k To Step Down The Voltage From 240v To 5v. The Out Of Voltage From The Current Sensor Is Compared With A Constant Voltage Value Of 3.8v, If The Output Voltage From The Current Sensor Is Within The Rated Scaled Value Of 3.8v To 5v; There Will Be An Output From The Comparison. The Output From Comparison Is Fed Into A Pi Controller. The Output From The Pi Controller Is Compared With A Triangular Wave Of 6v Amplitude To Generate Pwm Which Switches The Mosfet Of The Boost Converter. But If The Voltage From The Current Sensor Is Less Than 3.8v, There Will Be No Voltage Output Into The Pi Controller; No Comparison With The Triangular Wave And Therefore No Pwm Generated To Switch The Mosfet Of The Boost Converter. The Boost Converter Pwm Duty Cycle Is Determined By The Current Sensor Voltage To Give A Constant Output Voltage Of 415vdc.

Figure 11 Shows Three Waveforms A, B And C Form The Current/Voltage Controller Circuit. The Waveform A Displays The Output Waveform When The Current Sensor Voltage Is Greater Than 3.8v. The Waveform B Displays The Carrier Waveform (Triangular Waveform At 5khz) At 6v And Waveform C Displays The Comparing Of Waveform A With Waveform B To Generate Pwm Waveform.

Figure 12 Displays The Various Outputs Of The Boost Converter When There Is An Output From The Current/Voltage Circuit. Waveform A Is The Current With Value Of 5.9a, Waveform B Is The Voltage With Value
Three-Phase Five-Level Flying Capacitor Multilevelinverter For Harvesting Solar Power

Of 439.7v And Waveformc Is For Power With Value Of 2,612w.

Figure 10: Input Current/Voltage Controlled Boost Converter Circuit And Current, Voltage And Power Reading

Figure 11: PI output signal waveform voltage, triangular wave and the PWM form the

Figure 12: Waveforms Of The Output Current, Output Voltage And Output Power from the boost

Figure 13 Shows The Various Input Waveform Into The Three Phase Five Level Fc Mlii Controller Circuit Using Sinusoidal Pulse Width Modulated . The Three Sinewave Are Of 0, 120 And 240° And Four Carrier (Triangular) Waveforms At 0 To - 1; 1 To -2; 0 To -1 And -1 To -2 Voltage Levels. The Combined Waveform Is Inputted Into The Three-Phase Five-Level Fcmli.

Figure 14 Shows The Input Gating Signals Waveforms Of Phase 0. Waveform A Is The Waveform Generated When Sinewave Of 0 Cuts The First Carrier Waveform To Generate The Pwm Waveform A And Waveformb Is The Not Of Waveform A. Waveforms C, E, And G Are The Waveform When Sinewave Of 0 Cut The Second, Third And Fourth Carrier Shown In Figure 13 And Their Not Waveforms Are D, F And H Respectively. The Same Applies For The Other Phase Gating Signals But The Sine Waves Are At Different Phase Shift Of 120° And 240°.

Figure 15 Shows The Output Waveform Of The Three-Phase Five-Level Fcmli Using Sinusoidal Pulse Width Modulated. Where Waveformal Is For 0° Phase And Waveforms B1 And C1 Are Waveforms For 120° And 240° Phases Respectively. This Wave Can Be Filtered Into Pure Sine Wave By Using Filter Before Been Fed Into The Grid.

Figure 16 Shows That The Output Waveform Of The Three-Phase Five-Level Fcmli Voltage When A Load And No Filter Is Connected, When Filter Is Connected The Voltage Output Waveform Of The Same Fcmli Is Shown In Figure 17. While Figure 18 Shows The Current Output Waveform Of The Three-Phase Five-Level Fcmli With Load And No Filter. When Filter Is Connected Current Output Waveform Is Shown In Figure 19.
Figure 13: Input signals from the three sine wave generators and triangular wave generator.

Figure 14: Input of the gating signal waveform (output from the comparator waveform) for 0° phase.

Figure 15: Voltage output waveform of the three-phase five-level FCMLI with no filter and no load.

Figure 16: Voltage output waveform of the three-phase five-level FCMLI with no filter and with load.

Figure 18: Voltage output waveform of the three-phase five-level FCMLI with filter and with load.

**Figure 19:** Current output waveform of the three-phase five-level FCMLI with filter and with load.

Figure 20(A) to 20(C) Shows the Voltage THD of Phases 0° of the Three-Phase Five-Level; When No Load and No Filter Is Connected, When Load and No Filter Is Connected And When Load and Filter.
The Results It Shows That Filtered Voltage Has Lesser THD But There Is Voltage Drop Across The Filter.

![Figure 20: The THD Voltage Waveforms Of The Three-Phase Five Level Fcmli](image)

3.1 Discussion

The Output Of The Solar Pv From The Simulation Gives 240vdc, 15a Which Is Feed Into The Current/Voltage Controller Due To The Effect Of Irradiance On The Output Pv Solar Current. Table 2 Shows The Effect Of Irradiance On Current And Voltage And From The Table 2 It Shows That Irradiance Has More Effect On Current Than Voltage That Is Why A Current Controller Was Designed And Turning The PI, The Boost Converter Can Step Up The Voltage From 200-240vdc To A Constant 415vdc And This 450vdc Fed The Three-Phase Five-Level Fcmli And 415vac Is Generated At The Output Voltage Of The Three-Phase Five-Level Fcmli. Table 3 Shows The THD Of The Output Voltage, From The Table Its Shows That The Filtered Wave Have The Less THD Compared To Others.

### Table 2: Frequency Modulation Index Variation Table For Voltage Components On No Load

<table>
<thead>
<tr>
<th>Irradiance Values</th>
<th>Voltage Values @ 25 In Series</th>
<th>Current Values @ 25 In Parallel</th>
</tr>
</thead>
<tbody>
<tr>
<td>1000</td>
<td>249.9</td>
<td>17.24</td>
</tr>
<tr>
<td>750</td>
<td>245.9</td>
<td>12.93</td>
</tr>
<tr>
<td>500</td>
<td>242.0</td>
<td>8.63</td>
</tr>
<tr>
<td>250</td>
<td>232.5</td>
<td>4.31</td>
</tr>
<tr>
<td>100</td>
<td>210.3</td>
<td>1.73</td>
</tr>
<tr>
<td>50</td>
<td>128.7</td>
<td>0.86</td>
</tr>
<tr>
<td>10</td>
<td>25.96</td>
<td>0.17</td>
</tr>
</tbody>
</table>

### Table 3: Frequency Modulation Index Variation Table For Voltage Components On No Load

<table>
<thead>
<tr>
<th>State Of The Inverter</th>
<th>THD % Voltage Of Each State</th>
</tr>
</thead>
<tbody>
<tr>
<td>Connected Without Load And No Filter</td>
<td>33.97%</td>
</tr>
<tr>
<td>Connected With Load And No Filter</td>
<td>101.71%</td>
</tr>
<tr>
<td>Connected Without Load With Filter</td>
<td>16.76%</td>
</tr>
</tbody>
</table>
IV. Conclusion

In this project, the design of a three-phase five-level flying capacitor MLI for harvesting solar energy using SPWM technique for optimizes solar energy harvesting. The three-phase five-level flying capacitor MLI is a project design to help increase the grid power using renewable power source. The use of MLI for the system is able to easily achieve pure sinewave after filtering of the modified wave that comes out of the MLI. The pure sinewave can be connected to the grid to boost the national grid since the solar energy is free and readily available at the day time and where the usage of power energy is at its peak. It can also help to improve the poor power supply in developing countries and rural areas and can also be used as stand-alone power supply. Simulation of the system was done using MATLAB and results obtained were turned to obtain less THD for voltage waveform.

References